TEXAS INSTRUMENTS

ISO7142CC

ZHCSBM4B −SEPTEMBER 2013−REVISED AUGUST 2015

ISO7142CC 4242 V$_{PK}$ 小型封装低功耗四通道数字隔离器

1 特性

- 最大信号传输速率：50Mbps（5V 电源供电）
- 具有集成噪声滤波器的稳健设计
- 低功耗，每通道 I$_{CC}$ 典型值（3.3V 电源）：
  – 1Mbps 时为 1.3mA，25Mbps 时为 2.5mA
- 宽温度范围：–55°C 至 +125°C
- 典型值为 50kV/μs 的瞬态抗扰度
- 使用 SiO$_2$ 绝缘隔栅实现长使用寿命
- 可由 2.7V、3.3V 和 5V 电源供电
- 2.7V 和 5V 电平转换
- 小型四分之一尺寸小外形封装 (QSOP)-16 封装
- 安全及管理批准
  – 符合 UL 1577 标准且长达 1 分钟的 2500 V$_{RMS}$ 隔离
  – 符合 DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 标准的 4242 V$_{PK}$ 隔离
  – CSA 组件验收通知 5A，IEC 60950-1 和 IEC 61010-1 终端设备标准
  – 符合 GB4943.1-2011 的 CQC 认证

2 应用

- 通用隔离
- 工业自动化
- 电机控制
- 太阳能逆变器

3 说明

ISO7142CC 器件可提供符合 UL 1577 标准的长达 1 分钟且高达 2500 V$_{RMS}$ 的电流隔离，以及符合 VDE V 0884-10 标准的 4242 V$_{PK}$隔离。

ISO7142CC 是一款四通道隔离器，此隔离器具有两个正向和两个反向通道。此器件在由 5V 电源供电时的最大数据传输速率为 50Mbps，而在由 3.3V 或 2.7V 电源供电时的最大数据传输速率为 40Mbps。

ISO7142CC 器件的输入端集成有滤波器，适用于易受噪声干扰的应用。
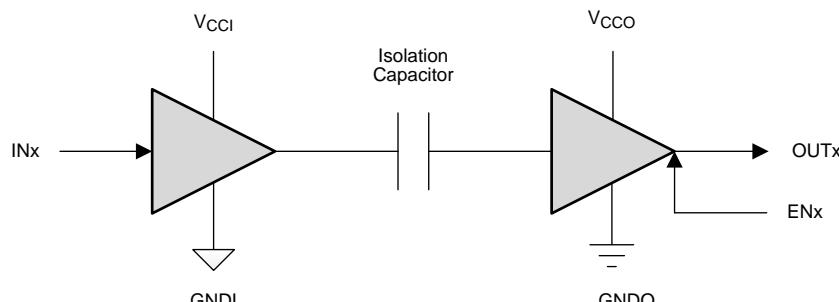
每个隔离通道都有一个由二氧化硅 (SiO$_2$) 绝缘隔栅分开的逻辑输入和输出缓冲器。与隔离式电源一起使用，这个器件可防止数据总线或者其它电路上的噪音电流进入本地接地和干扰或损坏敏感电路。该器件具有晶体管晶体管逻辑电路 (TTL) 输入阈值，并且可由 2.7V、3.3V 和 5V 电压供电运行。

器件信息[1]

器件型号	封装	封装尺寸（标称值）
ISO7142CC	SSOP (16)	4.90mm × 3.90mm

(1) 要了解所有可用封装，请见数据表末尾的可订购产品附录。

简化电路原理图

V$_{CCI}$ 和 GNDI 分别是输入通道的电源和接地连接。
V$_{CCO}$ 和 GNDO 分别是输出通道的电源和接地连接。

English Data Sheet: SLLSEF1

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4 修订历史记录

注：之前版本的页码可能与当前版本有所不同。

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5 Pin Configuration and Functions

**DBQ Package
16-Pin SSOP
Top View**

V_{CC1} 1	16 V_{CC2}

V_{CC1} — 1 ●   16 — V_{CC2}
$GND1$ — 2   15 — $GND2$
INA — 3   14 — $OUTA$
INB — 4   13 — $OUTB$
$OUTC$ — 5   12 — INC
$OUTD$ — 6   11 — IND
$EN1$ — 7   10 — $EN2$
$GND1$ — 8   9 — $GND2$

Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
EN1	7	I	Output enable 1. Output pins on side 1 are enabled when EN1 is high or open and in high-impedance state when EN1 is low.
EN2	10	I	Output enable 2. Output pins on side 2 are enabled when EN2 is high or open and in high-impedance state when EN2 is low.
GND1	2	—	Ground connection for V_{CC1}
	8		
GND2	9	—	Ground connection for V_{CC2}
	15		
INA	3	I	Input, channel A
INB	4	I	Input, channel B
INC	12	I	Input, channel C
IND	11	I	Input, channel D
OUTA	14	O	Output, channel A
OUTB	13	O	Output, channel B
OUTC	5	O	Output, channel C
OUTD	6	O	Output, channel D
V_{CC1}	1	—	Power supply, V_{CC1}
V_{CC2}	16	—	Power supply, V_{CC2}

3

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)[1]

			MIN	MAX	UNIT
	Supply voltage[2]	V_{CC1}, V_{CC2}	−0.5	6	V
	Voltage	INx, OUTx, ENx	−0.5	V_{CC} + 0.5[3]	V
I_O	Output current		−15	15	mA
T_J	Maximum junction temperature			150	°C
T_{stg}	Storage temperature		−65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
(3) Maximum voltage must not exceed 6 V.

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001[1]	±4000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101[2]	±1500	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

			MIN	NOM	MAX	UNIT
V_{CC1}, V_{CC2}	Supply voltage		2.7		5.5	V
I_{OH}	High-level output current	V_{CC} ≥ 3 V	−4			mA
		V_{CC} < 3 V	−2			
I_{OL}	Low-level output current				4	mA
V_{IH}	High-level input voltage		2		5.5	V
V_{IL}	Low-level input voltage		0		0.8	V
t_{ui}	Input pulse duration	V_{CC} ≥ 4.5 V	20			ns
		V_{CC} < 4.5 V	25			
$1 / t_{ui}$	Signaling rate	V_{CC} ≥ 4.5 V	0		50	Mbps
		V_{CC} < 4.5 V	0		40	
T_J	Junction temperature				136	°C
T_A	Ambient temperature		−55	25	125	°C

6.4 Thermal Information

		ISO7142CC	
	THERMAL METRIC[1]	DBQ (SSOP)	UNIT
		16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	104.5	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	57.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	46.8	°C/W
ψ_{JT}	Junction-to-top characterization parameter	18.3	°C/W
ψ_{JB}	Junction-to-board characterization parameter	46.4	°C/W
$R_{\theta JCbot}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, SPRA953.

6.5 Electrical Characteristics—5-V Supply

V_{CC1} and V_{CC2} at 5 V ± 10% (over recommended operating conditions unless otherwise noted.)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage	I_{OH} = −4 mA; see Figure 8	V_{CCO} [(1)] − 0.5				V
		I_{OH} = −20 µA; see Figure 8	V_{CCO} − 0.1				
V_{OL}	Low-level output voltage	I_{OL} = 4 mA; see Figure 8				0.4	V
		I_{OL} = 20 µA; see Figure 8				0.1	
$V_{I(HYS)}$	Input threshold voltage hysteresis				480		mV
I_{IH}	High-level input current	V_{IH} = V_{CCI}[(1)] at INx or ENx				10	µA
I_{IL}	Low-level input current	V_{IL} = 0 V at INx or ENx		−10			
CMTI	Common-mode transient immunity	V_I = V_{CCI} or 0 V; see Figure 11		25	70		kV/µs

(1) V_{CCI}= Supply voltage for the input channel; V_{CCO} = Supply voltage for the output channel

6.6 Supply Current Characteristics—5-V Supply

V_{CC1} and V_{CC2} at 5 V ± 10% (over recommended operating conditions unless otherwise noted.)

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT
Supply current for V_{CC1} and V_{CC2}	Disable	EN1 = EN2 = 0 V	I_{CC1}, I_{CC2}		0.8	1.6	mA
	DC to 1 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1} , I_{CC2}		3.3	5	
	10 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		4.9	7	
	25 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		7.3	10	
	50 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		11.1	14.5	

6.7 Electrical Characteristics—3.3-V Supply

V_{CC1} and V_{CC2} at 3.3 V ± 10% (over recommended operating conditions unless otherwise noted.)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage	I_{OH} = −4 mA; see Figure 8	V_{CCO} [(1)] − 0.5				V
		I_{OH} = −20 µA; see Figure 8	V_{CCO} − 0.1				
V_{OL}	Low-level output voltage	I_{OL} = 4 mA; see Figure 8				0.4	V
		I_{OL} = 20 µA; see Figure 8				0.1	
$V_{I(HYS)}$	Input threshold voltage hysteresis				460		mV
I_{IH}	High-level input current	V_{IH} = V_{CCI}[(1)] at INx or ENx				10	µA
I_{IL}	Low-level input current	V_{IL} = 0 V at INx or ENx		−10			
CMTI	Common-mode transient immunity	V_I = V_{CCI} or 0 V; see Figure 11		25	50		kV/µs

(1) V_{CCI}= Supply voltage for the input channel; V_{CCO} = Supply voltage for the output channel

6.8 Supply Current Characteristics—3.3-V Supply

V_{CC1} and V_{CC2} at 3.3 V ± 10% (over recommended operating conditions unless otherwise noted.)

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT
Supply current for V_{CC1} and V_{CC2}	Disable	EN1 = EN2 = 0 V	I_{CC1}, I_{CC2}		0.5	1	mA
	DC to 1 Mbps	DC signal: V_I = V_{CCI} or 0 V AC signal: All channels switching with square-wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		2.5	4	
	10 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		3.5	5	
	25 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		5	7	
	40 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		6.5	10	

6.9 Electrical Characteristics—2.7-V Supply

V_{CC1} and V_{CC2} at 2.7 V (over recommended operating conditions unless otherwise noted.)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage	I_{OH} = −2 mA; see Figure 8	V_{CCO} [(1)] − 0.3			V
		I_{OH} = −20 µA; see Figure 8	V_{CCO} − 0.1			
V_{OL}	Low-level output voltage	I_{OL} = 4 mA; see Figure 8			0.4	V
		I_{OL} = 20 µA; see Figure 8			0.1	
$V_{I(HYS)}$	Input threshold voltage hysteresis			360		mV
I_{IH}	High-level input current	V_{IH} = V_{CCI}[(1)] at INx or ENx			10	µA
I_{IL}	Low-level input current	V_{IL} = 0 V at INx or ENx	−10			
CMTI	Common-mode transient immunity	V_I = V_{CCI} or 0 V; see Figure 11	25	45		kV/µs

(1) V_{CCI}= Supply voltage for the input channel; V_{CCO} = Supply voltage for the output channel

6.10 Supply Current Characteristics—2.7-V Supply

V_{CC1} and V_{CC2} at 2.7 V (over recommended operating conditions unless otherwise noted.)

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT
Supply current for V_{CC1} and V_{CC2}	Disable	EN1 = EN2 = 0 V	I_{CC1}, I_{CC2}		0.4	0.8	mA
	DC to 1 Mbps	DC signal: V_I = V_{CCI} or 0 V AC signal: All channels switching with square-wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		2.2	3.5	
	10 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		3	4.2	
	25 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		4.2	5.5	
	40 Mbps	DC signal: V_I = V_{CCI} or 0 V, AC signal: All channels switching with square wave clock input; C_L = 15 pF	I_{CC1}, I_{CC2}		5.4	7.5	

6.11 Power Dissipation Characteristics

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
P_D	Device power dissipation	V_{CC1} = V_{CC2} = 5.5 V, T_J = 150°C, C_L = 15 pF Input a 25-MHz, 50% duty cycle square wave			170	mW

6.12 Switching Characteristics—5-V Supply

V_{CC1} and V_{CC2} at 5 V ± 10% (over recommended operating conditions unless otherwise noted.)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}, t_{PHL}	Propagation delay time	See Figure 8	15	21	38	ns
PWD[1]	Pulse width distortion $|t_{PHL} - t_{PLH}|$	See Figure 8			3.5	ns
$t_{sk(o)}$ [2]	Channel-to-channel output skew time	Same-direction channels			1.5	ns
		Opposite-direction channels			6.5	
$t_{sk(pp)}$ [3]	Part-to-part skew time				14	ns
t_r	Output signal rise time	See Figure 8		2.5		ns
t_f	Output signal fall time	See Figure 8		2.1		ns
t_{PHZ}, t_{PLZ}	Disable propagation delay, high/low-to-high impedance output	See Figure 9		7	12	ns
t_{PZH}	Enable propagation delay, high impedance-to-high output	See Figure 9		6	12	ns
t_{PZL}	Enable propagation delay, high impedance-to-low output	See Figure 9		12	23	us
t_{fs}	Fail-safe output delay time from input data or power loss	See Figure 10		8		µs
t_{GR}	Input glitch rejection time			9.5		ns

(1) Also known as pulse skew
(2) $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
(3) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals, and loads.

6.13 Switching Characteristics—3.3-V Supply

V_{CC1} and V_{CC2} at 3.3 V ± 10% (over recommended operating conditions unless otherwise noted.)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}, t_{PHL}	Propagation delay time	See Figure 8	16	25	46	ns
PWD[1]	Pulse width distortion $|t_{PHL} - t_{PLH}|$	See Figure 8			3	ns
$t_{sk(o)}$ [2]	Channel-to-channel output skew time	Same-direction Channels			2	ns
		Opposite-direction Channels			6.5	
$t_{sk(pp)}$ [3]	Part-to-part skew time				21	ns
t_r	Output signal rise time	See Figure 8		3		ns
t_f	Output signal fall time	See Figure 8		2.5		ns
t_{PHZ}, t_{PLZ}	Disable propagation delay, from high/low to high-impedance output	See Figure 9		9	14	ns
t_{PZH}	Enable propagation delay, from high-impedance to high output	See Figure 9		9	17	ns
t_{PZL}	Enable propagation delay, from high-impedance to low output	See Figure 9		12	24	us
t_{fs}	Fail-safe output delay time from input data or power loss	See Figure 10		7		µs
t_{GR}	Input glitch rejection time			11		ns

(1) Also known as pulse skew
(2) $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
(3) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

6.14 Switching Characteristics—2.7-V Supply

V_{CC1} and V_{CC2} at 2.7 V (over recommended operating conditions unless otherwise noted.)
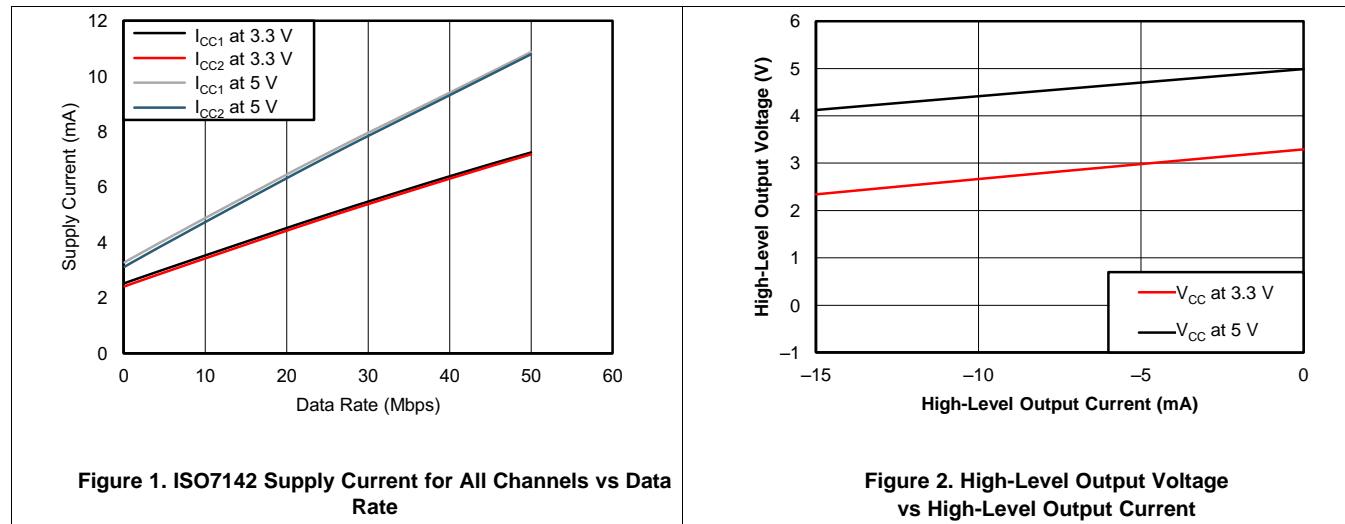
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}, t_{PHL}	Propagation delay time	See Figure 8	18	28	50	ns
PWD[1]	Pulse width distortion $|t_{PHL} - t_{PLH}|$	See Figure 8			3	ns
$t_{sk(o)}$ [2]	Channel-to-channel output skew time	Same-direction Channels			3	ns
		Opposite-direction Channels			8.5	ns
$t_{sk(pp)}$ [3]	Part-to-part skew time				24	ns
t_r	Output signal rise time	See Figure 8		3.5		ns
t_f	Output signal fall time	See Figure 8		2.8		ns
t_{PHZ}, t_{PLZ}	Disable propagation delay, from high/low to high-impedance output	See Figure 9		10	15	ns
t_{PZH}	Enable propagation delay, from high-impedance to high output	See Figure 9		10	19	ns
t_{PZL}	Enable propagation delay, from high-impedance to low output	See Figure 9		12	23	us
t_{fs}	Fail-safe output delay time from input data or power loss	See Figure 10		7		µs
t_{GR}	Input glitch rejection time			12		ns
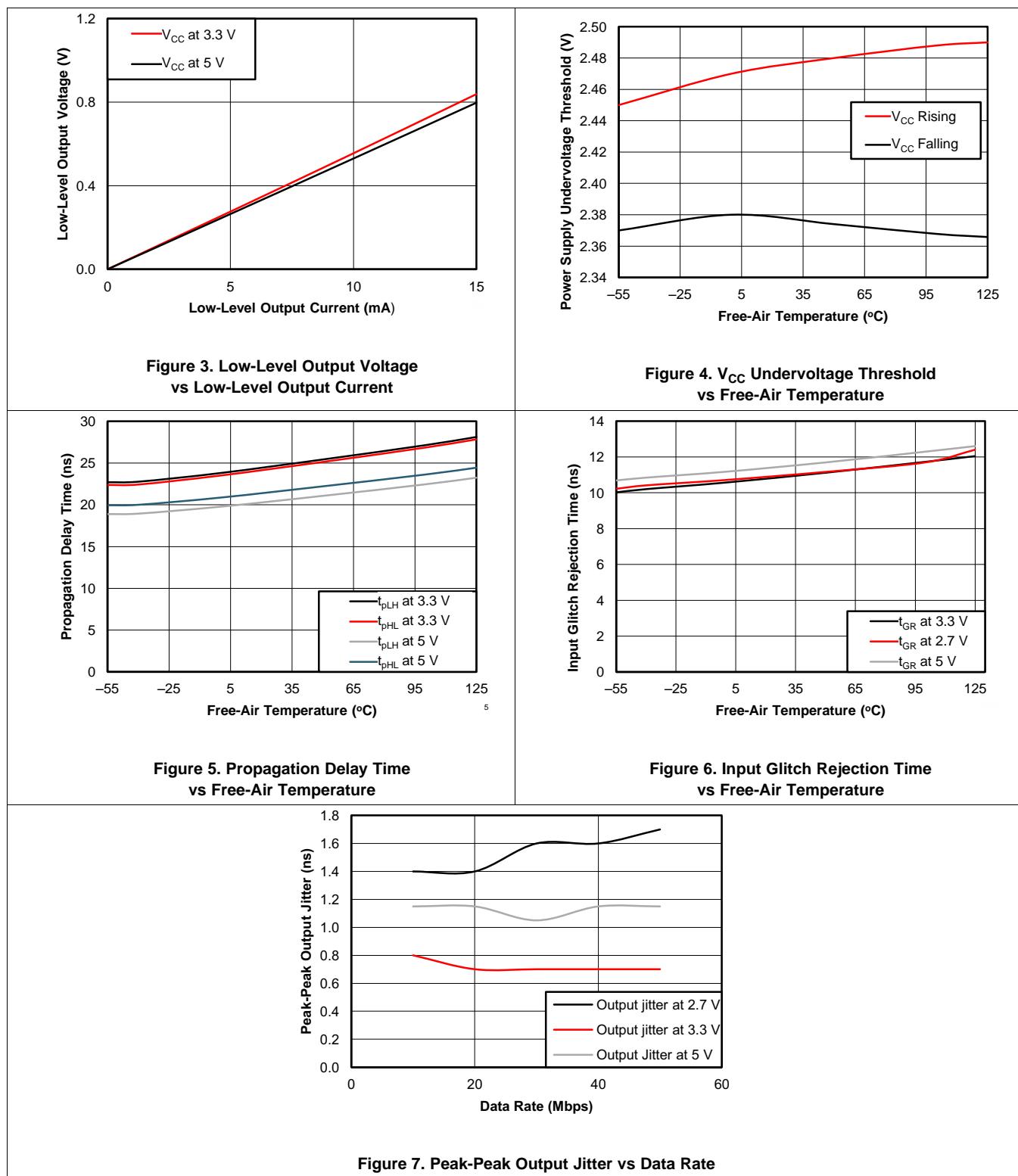
(1) Also known as pulse skew
(2) $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
(3) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals, and loads.

6.15 Typical Characteristics

Figure 1. ISO7142 Supply Current for All Channels vs Data Rate

Figure 2. High-Level Output Voltage vs High-Level Output Current
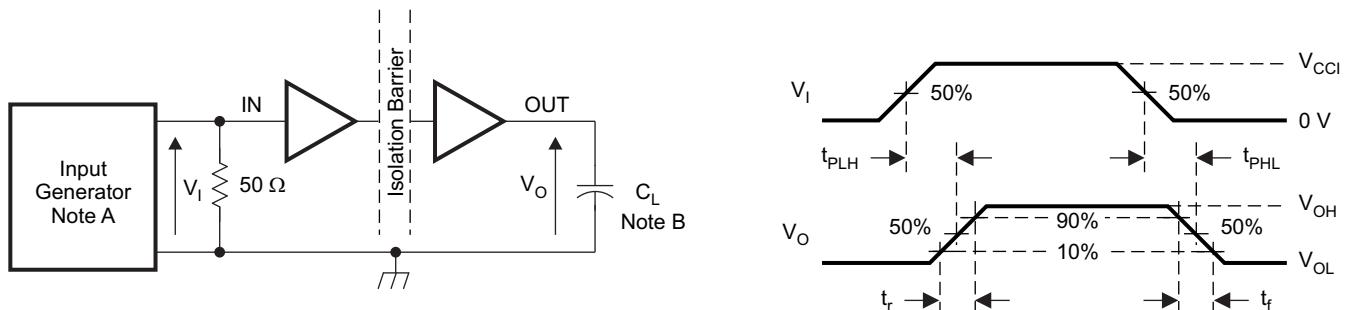
Typical Characteristics (continued)


Figure 3. Low-Level Output Voltage
vs Low-Level Output Current

Figure 4. V_{CC} Undervoltage Threshold
vs Free-Air Temperature

Figure 5. Propagation Delay Time
vs Free-Air Temperature

Figure 6. Input Glitch Rejection Time
vs Free-Air Temperature

Figure 7. Peak-Peak Output Jitter vs Data Rate
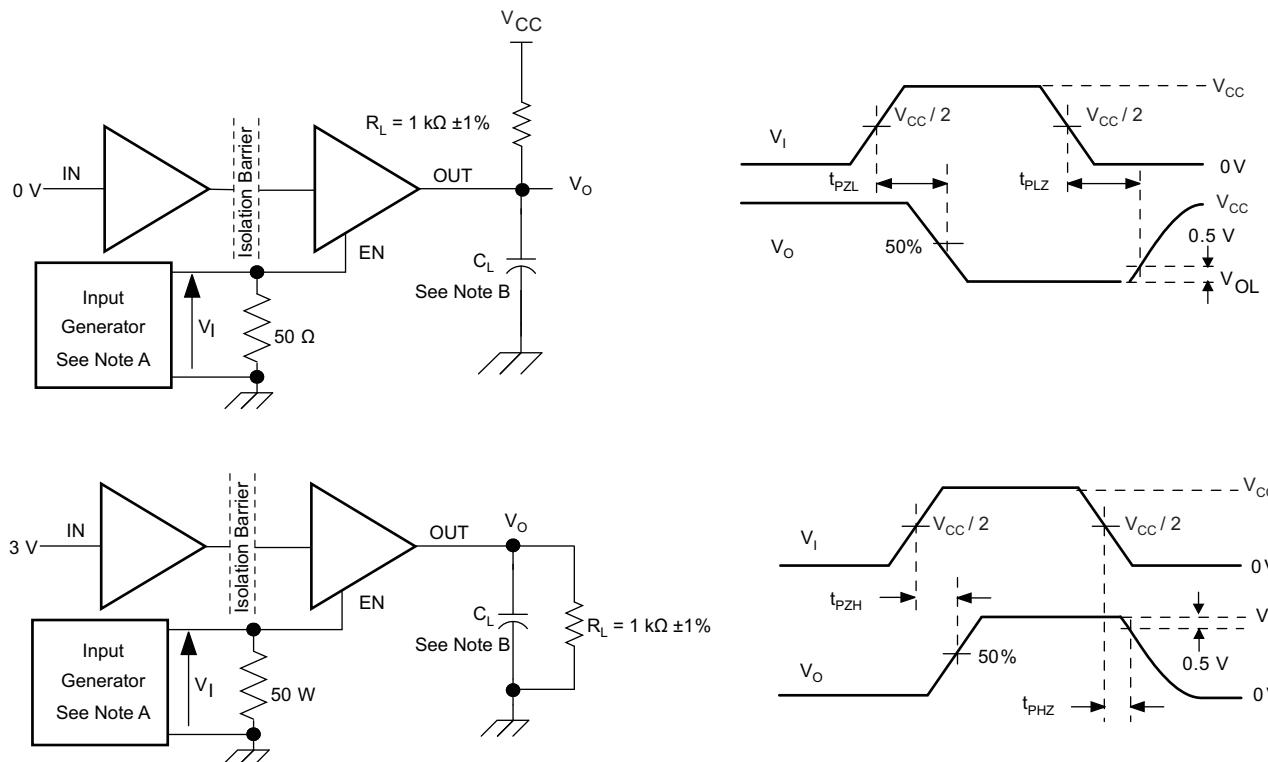
7 Parameter Measurement Information

A. The input pulse is supplied by a generator having the following characteristics: PRR ≤ 50 kHz, 50% duty cycle, t_r ≤ 3 ns, t_f ≤ 3 ns, Z_O = 50 Ω. At the input, a 50-Ω resistor is required to terminate the input-generator signal. It is not needed in an actual application.

B. C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.
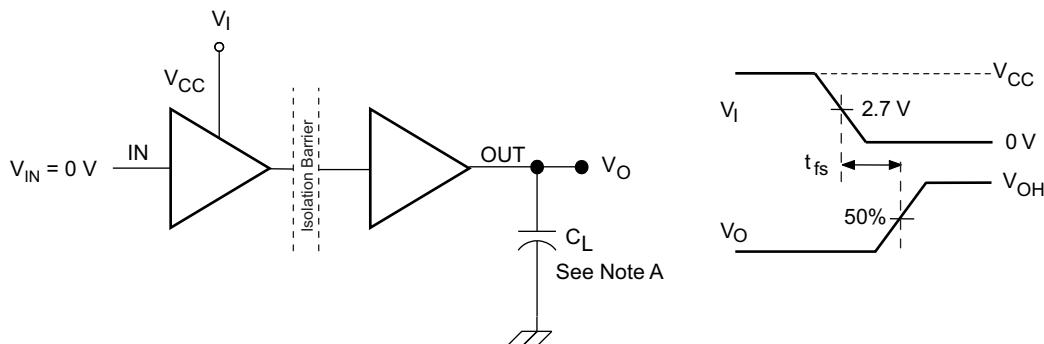
**Figure 8. Switching-Characteristics Test Circuit and Voltage Waveforms**

A. The input pulse is supplied by a generator having the following characteristics: PRR ≤ 50 kHz, 50% duty cycle, t_r ≤ 3 ns, t_f ≤ 3 ns, Z_O = 50 Ω.

B. C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.
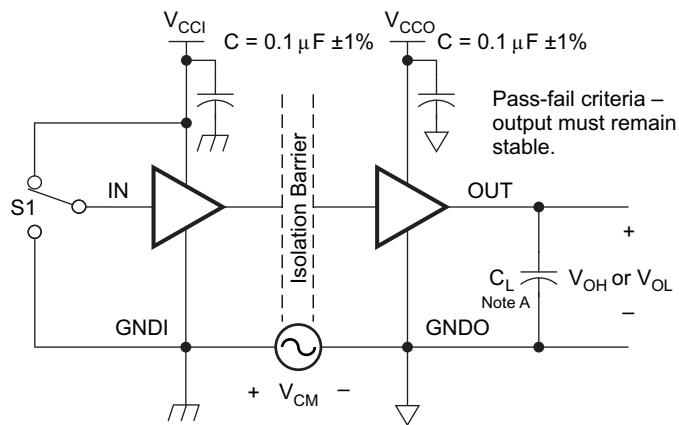
Figure 9. Enable/Disable Propagation Delay-Time Test Circuit and Waveform

Parameter Measurement Information (continued)

A. C_L = 15 pF and includes instrumentation and fixture capacitance within ±20%.

Figure 10. Failsafe Delay-Time Test Circuit and Voltage Waveforms

A. C_L = 15pF and includes instrumentation and fixture capacitance within ±20%.

Figure 11. Common-Mode Transient Immunity Test Circuit

8 Detailed Description

8.1 Overview

The isolator in Figure 12 is based on a capacitive isolation barrier technique. The I/O channel of the device consists of two internal data channels, a high-frequency channel (HF) with a bandwidth from 100 kbps up to 50 Mbps, and a low-frequency channel (LF) covering the range from 100 kbps down to DC. In principle, a single-ended input signal entering the HF-channel is split into a differential signal through the inverter gate at the input. The following capacitor-resistor networks differentiate the signal into transients, which then are converted into differential pulses by two comparators. The comparator outputs drive a NOR-gate flip-flop whose output feeds an output multiplexer. A decision logic (DCL) at the driving output of the flip-flop measures the durations between signal transients. If the duration between two consecutive transients exceeds a certain time limit, (as in the case of a low-frequency signal), the DCL forces the output-multiplexer to switch from the high- to the low-frequency channel.

Because low-frequency input signals require the internal capacitors to assume prohibitively large values, these signals are pulse-width modulated (PWM) with the carrier frequency of an internal oscillator, thus creating a sufficiently high frequency signal, capable of passing the capacitive barrier. As the input is modulated, a low-pass filter (LPF) is needed to remove the high-frequency carrier from the actual data before passing it on to the output multiplexer.

8.2 Functional Block Diagram
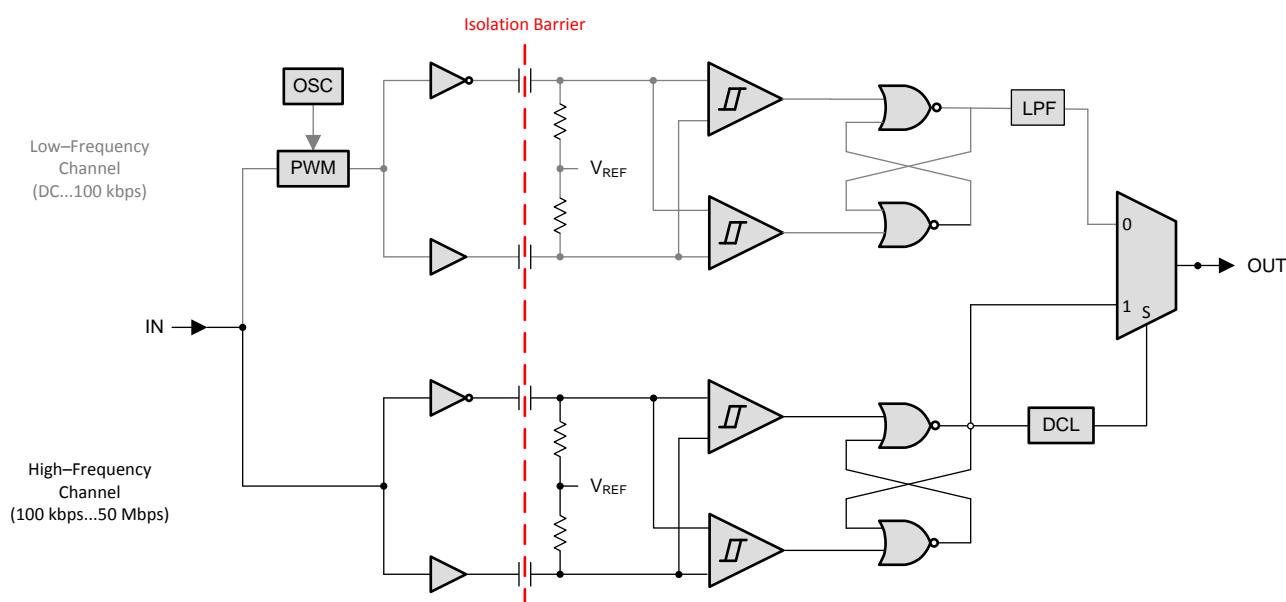
Figure 12. Conceptual Block Diagram of a Digital Capacitive Isolator

8.3 Feature Description

8.3.1 Insulation and Safety-Related Specifications

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	0.014			mm
C_I [1]	Input capacitance	$V_I = V_{CC}/2 + 0.4 \sin(2\pi ft)$, f = 1 MHz, V_{CC} = 5 V		2		pF
DIN V VDE V 0884-10 (VDE V 0884-10):2006-12						
V_{IOTM}	Maximum transient isolation voltage				4242	V_{PK}
V_{IORM}	Maximum working isolation voltage				566	V_{PK}
V_{PR}	Input-to-output test voltage	After Input/Output safety test subgroup 2/3, $V_{PR} = V_{IORM}$ x 1.2, t = 10 s, Partial discharge < 5 pC			679	V_{PK}
		Method a, After environmental tests subgroup 1, $V_{PR} = V_{IORM}$ x 1.6, t = 10 s, Partial discharge < 5 pC			906	
		Method b1, 100% production test, $V_{PR} = V_{IORM}$ x 1.875, t = 1 s, Partial discharge < 5 pC			1061	
L(I01)	Minimum air gap (clearance)	Shortest terminal to terminal distance through air	3.7			mm
L(I02)	Minimum external tracking (creepage)	Shortest terminal to terminal distance across the package surface	3.7			mm
	Pollution degree			2		
CTI	Tracking resistance (comparative tracking index)	DIN EN 60112 (VDE 0303-11); IEC 60112	≥400			V
R_{IO} [2]	Isolation resistance, input to output	V_{IO} = 500 V, T_A = 25°C		>10^{12}		Ω
		V_{IO} = 500 V, 100°C ≤ T_A ≤ 125°C		>10^{11}		
		V_{IO} = 500 V, T_S = 150°C		>10^9		
C_{IO} [2]	Barrier capacitance, input to output	$V_I = 0.4 \sin(2\pi ft)$, f = 1 MHz		2.4		pF
UL 1577						
V_{ISO}	Withstanding Isolation voltage	$V_{TEST} = V_{ISO}$= 2500 V_{RMS}, 60 sec (qualification); $V_{TEST} = 1.2 * V_{ISO}$= 3000 V_{RMS}, 1 sec (100% production)			2500	V_{RMS}

(1) Measured from input data pin to ground.
(2) All pins on each side of the barrier tied together creating a two-terminal device.

> **NOTE**
> Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed circuit board do not reduce this distance.
>
> Creepage and clearance on a printed circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.

Table 1. IEC 60664-1 Ratings Table

PARAMETER	TEST CONDITIONS	SPECIFICATION
Material Group		II
Installation classification / Overvoltage Category for Basic Insulation	Rated mains voltage ≤ 150 V_{RMS}	I–IV
	Rated mains voltage ≤ 300 V_{RMS}	I–III

8.3.2 Regulatory Information

VDE	UL	CSA	CQC
Certified according to DIN V VDE V 0884-10 (VDE V 0884-10):2006-12 and DIN EN 61010-1 (VDE 0411-1):2011-07	Certified under UL 1577 Component Recognition Program	Approved under CSA Component Acceptance Notice 5A, IEC 60950-1 and IEC 61010-1	Certified according to GB 4943.1-2011
Basic Insulation; Maximum transient Isolation IsolatiIsolationvoltage, 4242 V_{PK} Maximum working isolation voltage, 566 V_{PK}	Single protection, 2500 V_{RMS} [(1)]	3000 V_{RMS} Isolation rating; 185 V_{RMS} Reinforced Insulation and 370 V_{RMS} Basic Insulation per CSA 60950-1-07+A1+A2 and IEC 60950-1 2nd Ed.+A1+A2; 150 V_{RMS} Reinforced Insulation and 300 V_{RMS} Basic Insulation per CSA 61010-1-12 and IEC 61010-1 3rd Ed.	Basic Insulation, Altitude ≤ 5000m, Tropical climate, 250 V_{RMS} maximum working voltage.
File number: 40016131	File number: E181974	Master contract number: 220991	Certificate number: CQC14001109540

(1)   Production tested ≥ 3000 V_{RMS} for 1 second in accordance with UL 1577.

8.3.3 Safety Limiting Values

Safety limiting intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the IO can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to overheat the die and damage the isolation barrier, potentially leading to secondary system failures.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_S	Safety input, output, or supply current	θ_{JA} = 104.5°C/W, V_I = 5.5 V, T_J = 150°C, T_A = 25°C			217	mA
		θ_{JA} = 104.5°C/W, V_I = 3.6 V, T_J = 150°C, T_A = 25°C			332	
	DBQ-16	θ_{JA} = 104.5°C/W, V_I = 2.7 V, T_J = 150°C, T_A = 25°C			443	
T_S	Maximum safety temperature				150	°C

The safety-limiting constraint is the absolute-maximum junction temperature specified in the *Absolute Maximum Ratings* table. The power dissipation and junction-to-air thermal impedance of the device installed in the application hardware determines the junction temperature. The assumed junction-to-air thermal resistance in the *Thermal Information* table is that of a device installed on a high-K test board for leaded surface-mount packages. The power is the recommended maximum input voltage times the current. The junction temperature is then the ambient temperature plus the power times the junction-to-air thermal resistance.
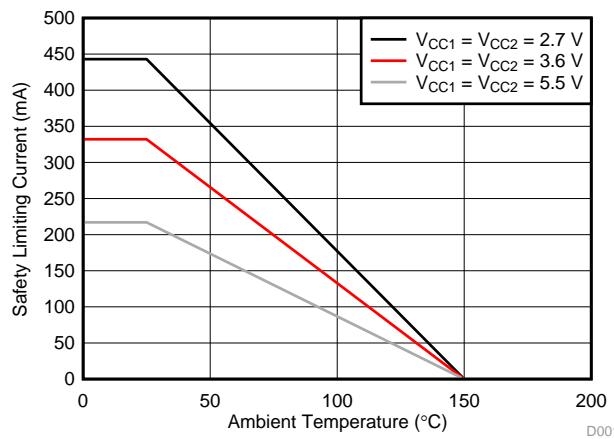

Figure 13. Thermal Derating Curve for Safety Limiting Current per VDE

8.4 Device Functional Modes

Table 2 lists the functional modes for the ISO7142CC.

Table 2. Function Table[1]

V_CCI	V_CCO	INPUT (INx)	OUTPUT ENABLE (ENx)	OUTPUT (OUTx)
PU	PU	H	H or open	H
		L	H or open	L
		X	L	Z
		Open	H or open	H
PD	PU	X	H or open	H
PD	PU	X	L	Z
X	PD	X	X	Undetermined

(1) V_{CCI} = Input-side Supply Voltage; V_{CCO} = Output-side Supply Voltage; PU = Powered Up ($V_{CC} \geq 2.7$ V); PD = Powered Down ($V_{CC} \leq 2.1$ V); X = Irrelevant; H = High Level; L = Low Level; Z = High Impedance

### 8.4.1 Device I/O Schematics
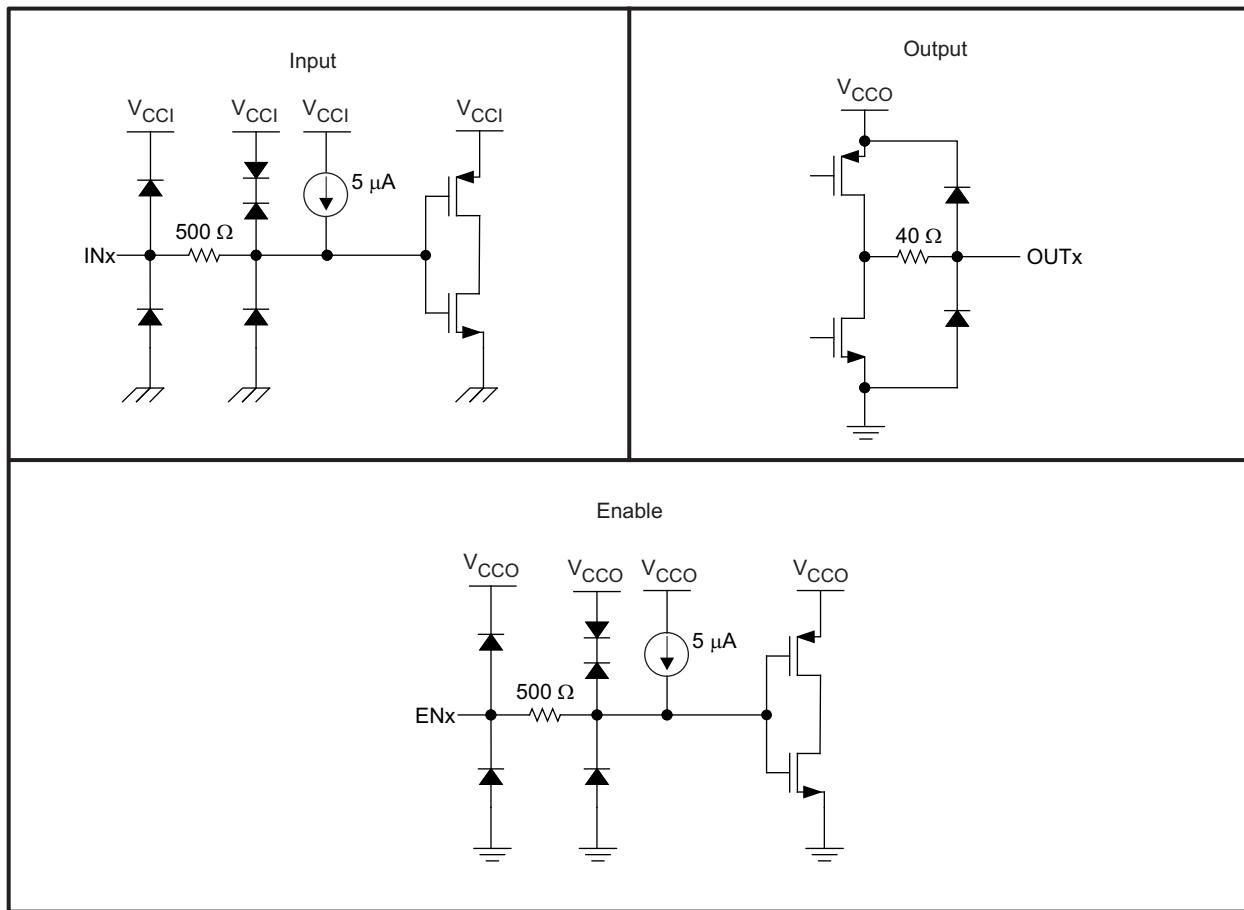
Figure 14. Device I/O Schematics

9 Application and Implementation

> **NOTE**
>
> Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The ISO7142CC device uses single-ended TTL-logic switching technology. The supply voltage range is from 2.7 V to 5.5 V for both supplies, V_{CC1} and V_{CC2}. When designing with digital isolators, keep in mind that because of the single-ended design structure the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is, μC or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

9.2 Typical Application

Figure 15 shows the typical isolated RS-232 interface implementation.
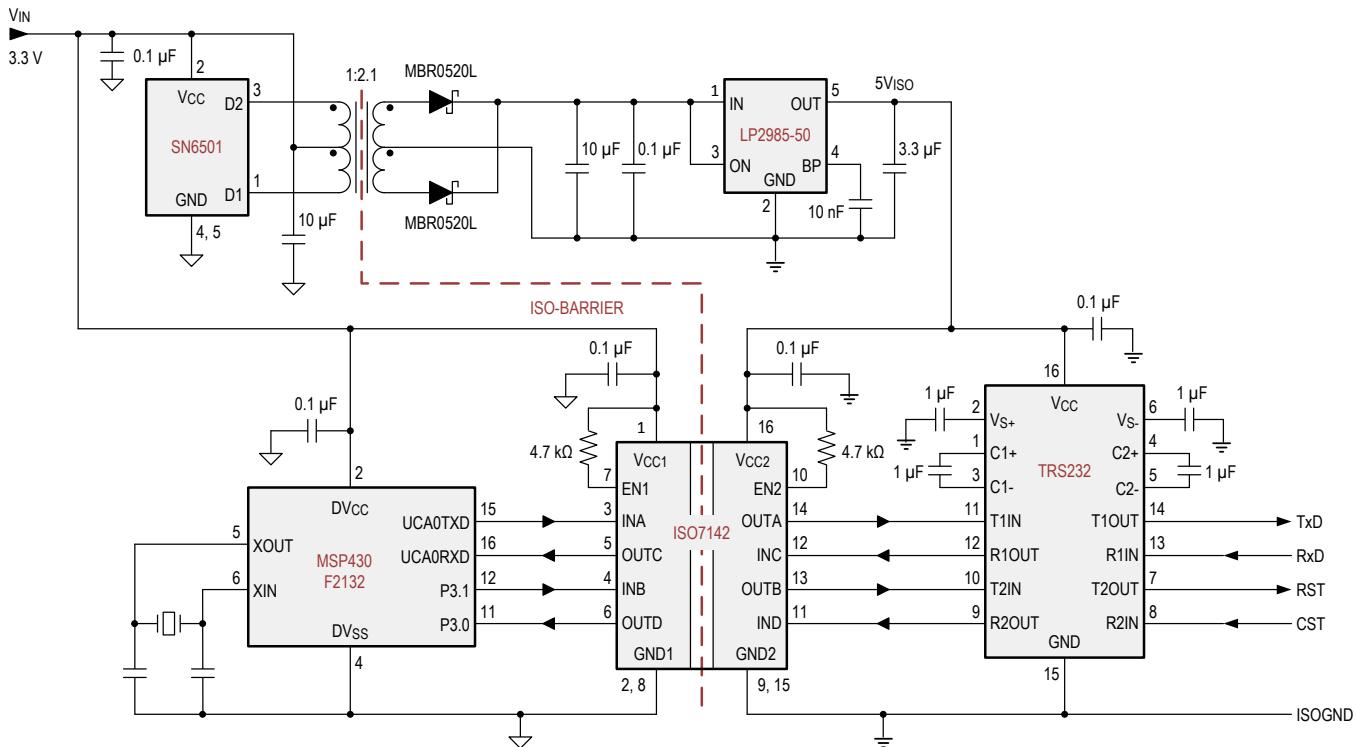
Figure 15. Typical Isolated RS-232 Application Circuit for ISO7142

9.2.1 Design Requirements

Unlike optocouplers, which require external components to improve performance, provide bias, or limit current, the ISO7142CC device only requires two external bypass capacitors to operate.

Typical Application (continued)

9.2.2 Detailed Design Procedure

Figure 16 shows the hookup of a typical ISO7142CC circuit. The only external components are two bypass capacitors.
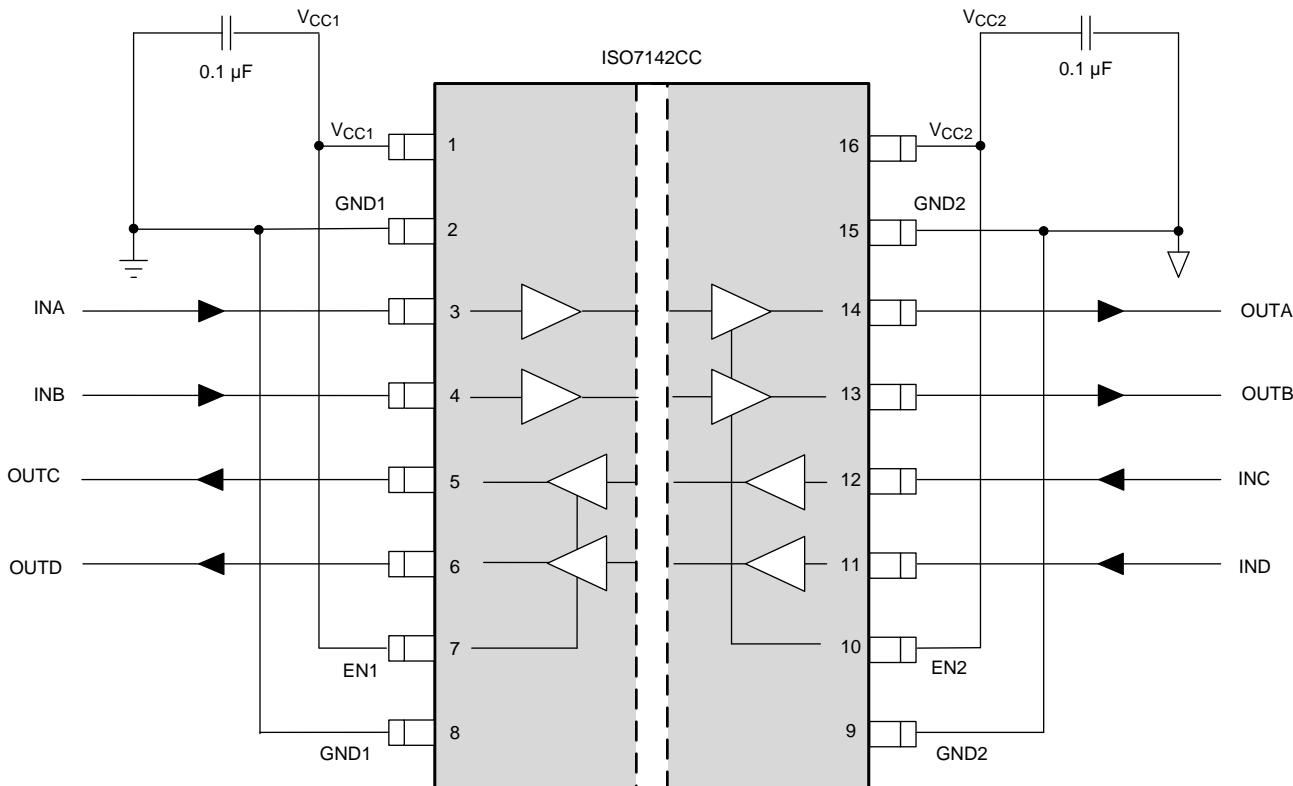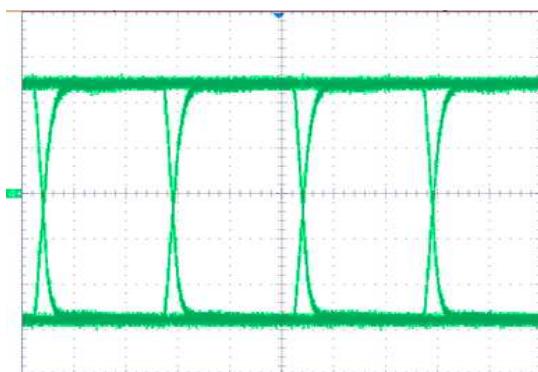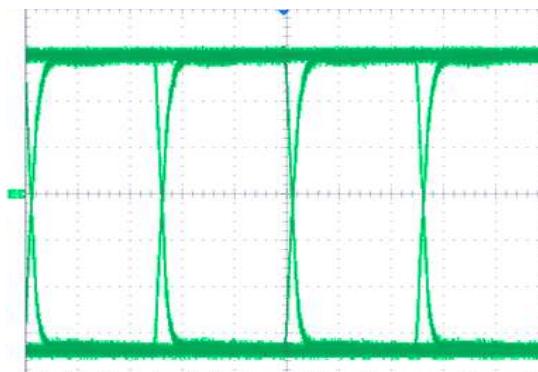
Figure 16. Typical ISO7142CC Circuit Hook-up

9.2.3 Application Curves
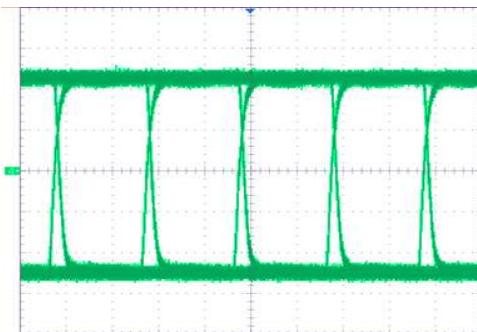
Figure 17. Typical Eye Diagram at 40 Mbps, PRBS 2^{16} - 1, 2.7-V Operation

Figure 18. Typical Eye Diagram at 40 Mbps, PRBS 2^{16} - 1, 3.3-V Operation

Typical Application (continued)

**Figure 19. Typical Eye Diagram at 50 Mbps,
PRBS 2^{16} - 1, 5-V Operation**

10 Power Supply Recommendations

To help ensure reliable operation supply voltages, a 0.1-µF bypass capacitor is recommended at the input and output supply pins (V_{CC1} and V_{CC2}). The capacitors should be placed as close to the supply pins as possible. If only a single primary-side power supply is available in an application, isolated power can be generated for the secondary-side with the help of a transformer driver such as Texas Instruments' SN6501-Q1. For such applications, detailed power supply design and transformer selection recommendations are available in SN6501-Q1 datasheet (SLLSEF3).

11 Layout

11.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see Figure 20). Layer stacking should be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of their inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100 pF/in^2.
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

If an additional supply voltage plane or signal layer is needed, add a second power and ground plane system to the stack to keep it symmetrical. This makes the stack mechanically stable and prevents it from warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, see the application note, *Digital Isolator Design Guide*, SLLA284.

Layout Guidelines (continued)

11.1.1 PCB Material

For digital circuit boards operating below 150 Mbps, (or rise and fall times higher than 1 ns), and trace lengths of up to 10 inches, use standard FR-4 UL 94 V-0 printed circuit board. This PCB is preferred over cheaper alternatives because of lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and self-extinguishing flammability-characteristics.

11.2 Layout Example

High-speed traces

10 mils

Ground plane

40 mils

Keep this space free from planes, traces, pads, and vias

FR-4
$\varepsilon_r \sim 4.5$

Power plane

10 mils

Low-speed traces
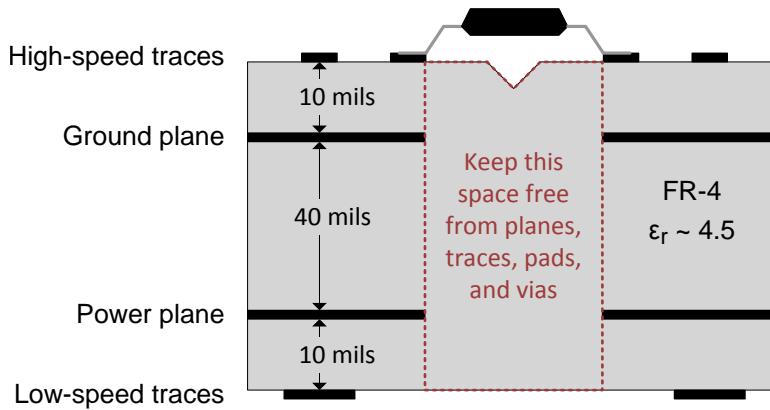
Figure 20. Recommended Layer Stack

12 器件和文档支持

12.1 文档支持

12.1.1 相关文档

相关文档如下：

- 《数字隔离器设计指南》，SLLA284
- 《隔离相关术语》，SLLA353
- 《ISO71xx EVM 用户指南》，SLLU179
- 《LP2985 具有关断模式的 150mA 低噪声低压降滤波器》，SLVS522
- 《SN6501 用于隔离电源的变压器驱动器》，SLLSEA0
- 《MSP430F2132 混合信号微控制器》，SLAS578
- 《TRS232 具有 IEC61000-4-2 保护的双路 RS-232 驱动器/接收器》，SLLS861

12.2 社区资源

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community *TI's Engineer-to-Engineer (E2E) Community.* Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

12.3 商标

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12.4 静电放电警告

这些装置包含有限的内置 ESD 保护。 存储或装卸时，应将导线一起截短或将装置放置于导电泡棉中，以防止 MOS 门极遭受静电损伤。

12.5 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 机械、封装和可订购信息

以下页中包括机械、封装和可订购信息。这些信息是针对指定器件可提供的最新数据。这些数据会在无通知且不对本文档进行修订的情况下发生改变。欲获得该数据表的浏览器版本，请查阅左侧的导航栏。

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ISO7142CCDBQ	ACTIVE	SSOP	DBQ	16	75	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7142C	Samples
ISO7142CCDBQR	ACTIVE	SSOP	DBQ	16	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-55 to 125	7142C	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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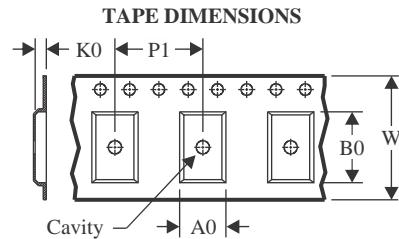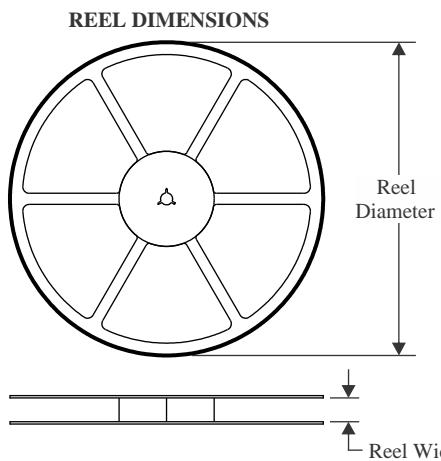
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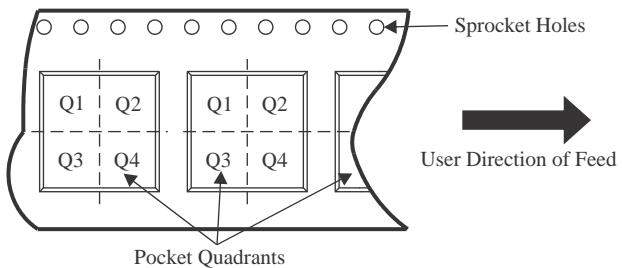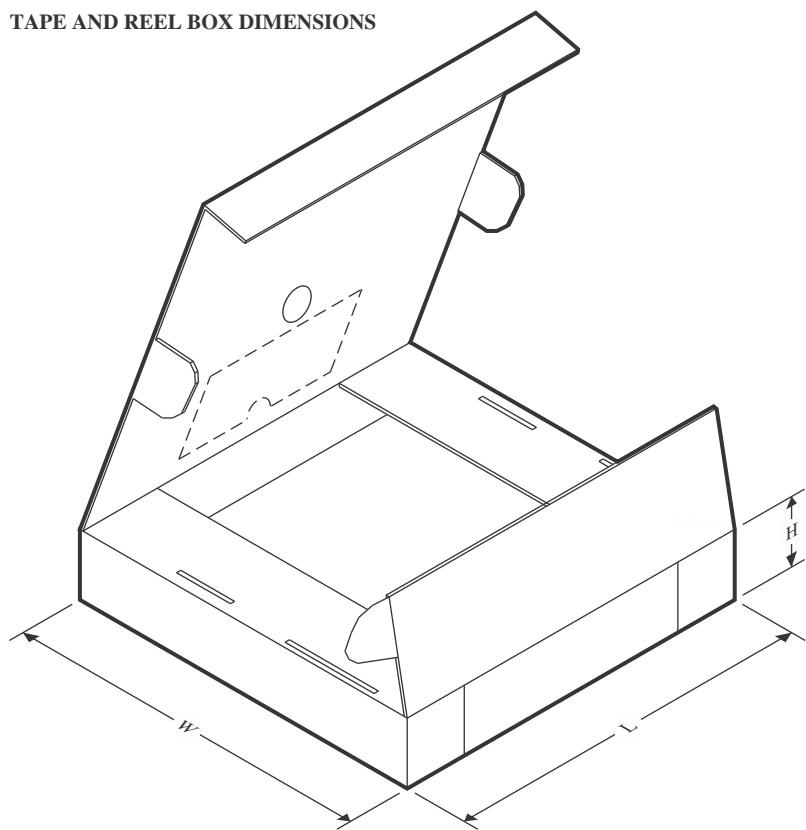
TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSIONS

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

User Direction of Feed
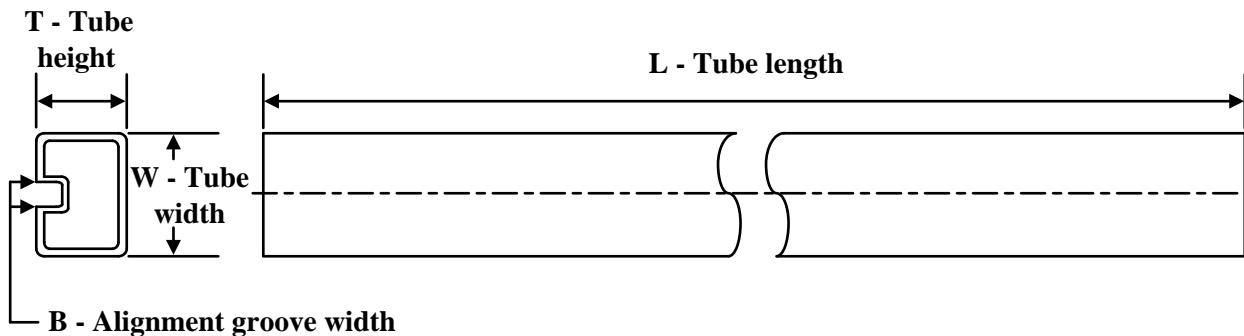
*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ISO7142CCDBQR	SSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

PACKAGE MATERIALS INFORMATION

TAPE AND REEL BOX DIMENSIONS

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ISO7142CCDBQR	SSOP	DBQ	16	2500	350.0	350.0	43.0

TUBE

T - Tube height

L - Tube length

W - Tube width

B - Alignment groove width

*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
ISO7142CCDBQ	DBQ	SSOP	16	75	505.46	6.76	3810	4
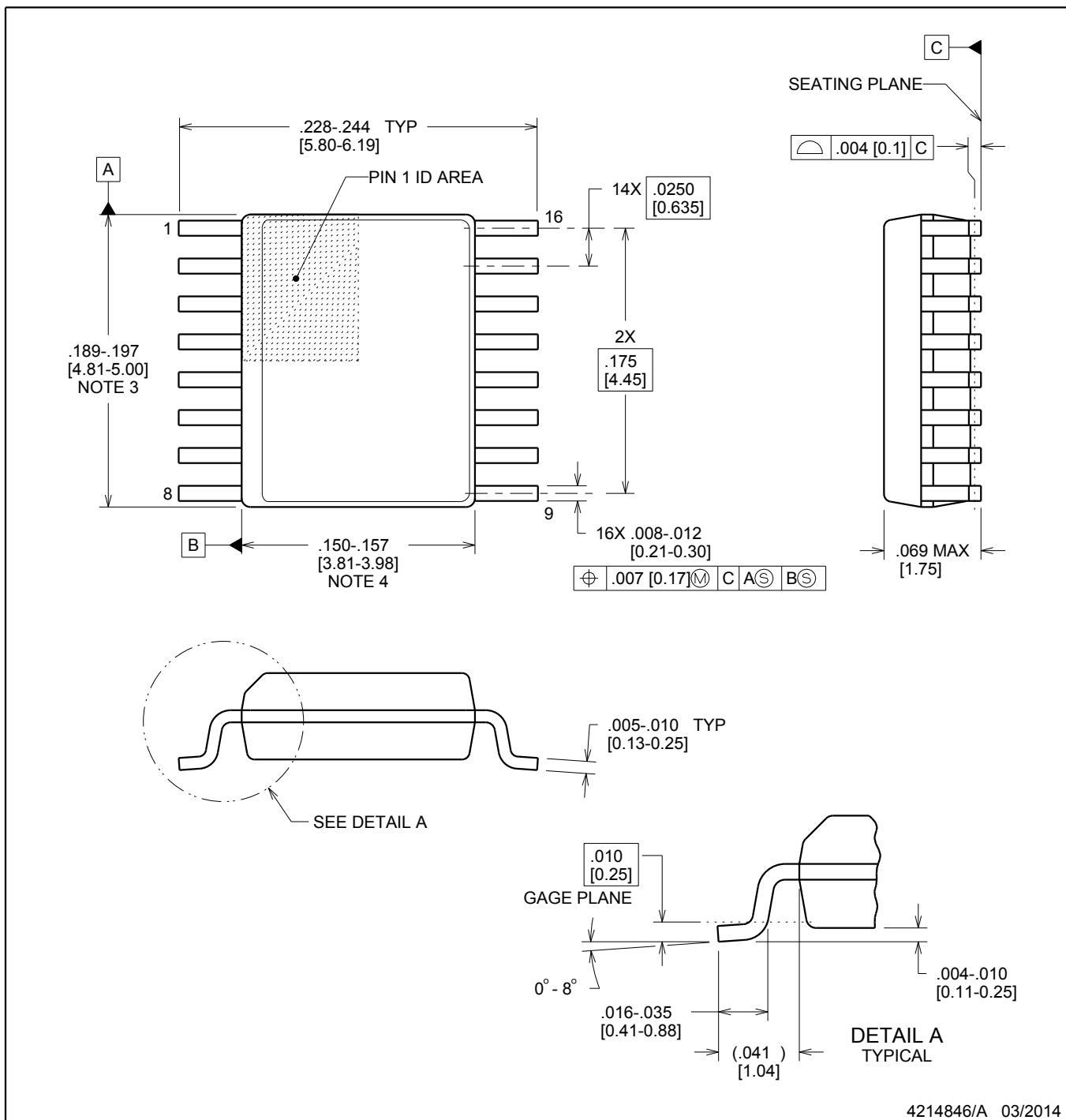
.228-.244 TYP
[5.80-6.19]

A

PIN 1 ID AREA

14X .0250
[0.635]

1

16

2X
.175
[4.45]

.189-.197
[4.81-5.00]
NOTE 3

8

9

B

.150-.157
[3.81-3.98]
NOTE 4

16X .008-.012
[0.21-0.30]

⊕ .007 [0.17] Ⓜ C AⓈ BⓈ

C

SEATING PLANE

◠ .004 [0.1] C

.069 MAX
[1.75]

SEE DETAIL A

.005-.010 TYP
[0.13-0.25]

.010
[0.25]
GAGE PLANE

0° - 8°

.016-.035
[0.41-0.88]

(.041)
[1.04]

.004-.010
[0.11-0.25]

DETAIL A
TYPICAL

4214846/A   03/2014

NOTES:
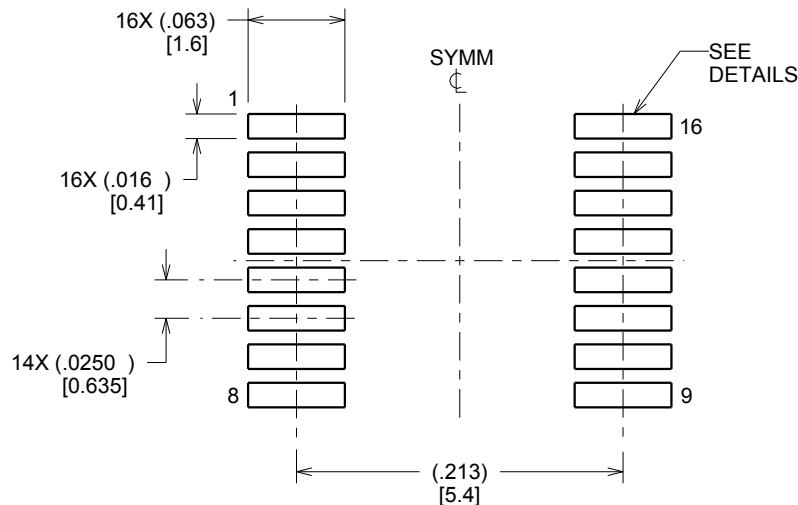
1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches.
   Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
   exceed .006 inch, per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MO-137, variation AB.

DBQ0016A

16X (.063)
[1.6]

SYMM
℄

SEE
DETAILS

1

16

16X (.016)
[0.41]

14X (.0250)
[0.635]

8

9

(.213)
[5.4]

LAND PATTERN EXAMPLE
SCALE:8X

METAL

SOLDER MASK
OPENING

SOLDER MASK
OPENING

METAL

.002 MAX
[0.05]
ALL AROUND

.002 MIN
[0.05]
ALL AROUND

NON SOLDER MASK
DEFINED

SOLDER MASK
DEFINED

SOLDER MASK DETAILS

4214846/A   03/2014

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
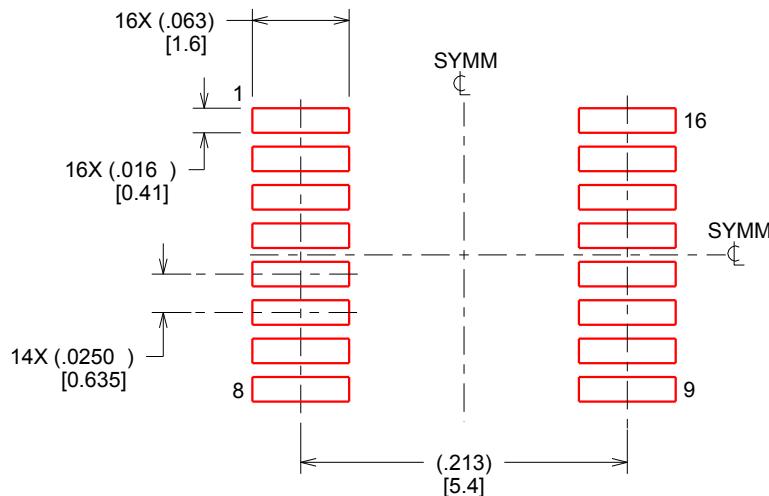
DBQ0016A

SSOP - 1.75 mm max height

16X (.063)
[1.6]

SYMM
₵

1

16X (.016)
[0.41]

16

SYMM
₵

14X (.0250)
[0.635]

8

9

(.213)
[5.4]

SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.127 MM] THICK STENCIL
SCALE:8X

4214846/A 03/2014

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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